Blind Source Separation Coping with the Change of the Number of Sources

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Abstract. This manuscript deals with the blind source separation problem with an instantaneous but dynamical mixture model. This study is limited to the case when the number of sources is time-variant. Theoretically, when new sources are detected, a new separating matrix should be estimated in order to extract all sources. However this effort implies an overwhelm computational cost. Our idea consists to use the previous separating matrix which was estimated before the appearance of the new sources. Owing to this point, the computational time and cost can be effectively reduced compared with the conventional separation scheme. Our new algorithm was corroborated with many simulations. Some results are given in the manuscript. The obtained and presented results clearly show that the proposed method outperformed the conventional method in processing time as well as in separation quality.

Keywords: blind source separation, time-variant system, dynamical instantaneous mixtures, independent component analysis.

1 Introduction

In the last decade, independent component analysis (ICA) has been greatly developed [1,2]. ICA is often used to solve blind source separation (BSS) problems, which means the estimation of original sources from their mixtures, using the only observed signals without any knowledge about the sources and the mixing process. Applying ICA algorithms the original sources can be estimated up to scaling and permutation factors.

A number of researchers have been struggling with BSS problems. However, most of the researches on BSS assume the time-invariant systems. The number of sources is a very important factor for BSS problems. However, most of the algorithms assumes that the number of sources is constant and known a priori. Ye et al. have proposed the BSS without knowing the number of sources [3]. But they



Fig. 1. Mixing and separating process. (a) The process before the appearance of $s_2(t)$. Only $s_1(t)$ exists. (b) The process after the appearance of $s_2(t)$.

have not considered the change of the number. For practical use, we need a solution of the BSS problems in time-variant systems, where the properties of the mixing process and sources may change. Some researchers have proposed methods to solve the blind separation of moving sources [4,5]. When source moves, the property of the mixing process varies, but the dimensions of the mixing matrix and the sources are constant. Therefore we deal with the case where the number of sources is not constant, that is, the dimensions of the mixing matrix and the sources are variable. Until now, such a problem has not been solved. In particular we consider the case that the number of sources increases because a decrease of sources does not affect the subsequent separation process.

In this manuscript, in order to solve the BSS problem with the change of the number of sources, the following conditions are assumed;

- A1) The mixing process is invertible and instantaneous.
- A2) There are no noise.
- A3) The number of active sources is always known.
- A4) The signals are real-valued.

Hereafter, the scheme of the separation coping with the case that new sources appear is proposed. Making use of the covariance matrix of the mixed signals, a new matrix which blocks undesired sources can be estimated. Through computer simulations, it is clarified that the proposed method can work efficiently.

2 Problem Formulation

It is supposed that n_1 sources initially exist for $t < T_0$ and at time $t = T_0$ new n_2 sources suddenly appear, i.e., $(n_1 + n_2)$ sources are active for the period $t \ge T_0$.

The former sources are denoted as $s_1(t)$, the latter source $s_2(t)$. We call $s_1(t)$ initial sources and $s_2(t)$ additional sources. We unite them into s(t):

$$\boldsymbol{s}(t) = \begin{bmatrix} \boldsymbol{s}_1(t) \\ \boldsymbol{s}_2(t) \end{bmatrix}. \tag{1}$$

Their mixtures are observed by multiple sensors whose number m is greater than or equal to that of whole sources $(n_1 + n_2)$. Mixed signals are denoted by $\boldsymbol{x}_1(t)$ $(t < T_0)$ or $\boldsymbol{x}_2(t)$ $(t \ge T_0)$. The initial mixed signals $\boldsymbol{x}_1(t)$ can be expressed as follows (See Fig. 1 (a).):

$$\boldsymbol{x}_{1}(t) = [\boldsymbol{H}_{1} \ \boldsymbol{H}_{2}] \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \boldsymbol{0}_{n_{2}} \end{bmatrix}.$$
(2)

 $\boldsymbol{x}_1(t)$ can be rewritten in another way:

$$\boldsymbol{x}_{1}(t) = [\boldsymbol{H}_{1} \ \boldsymbol{O}_{m \times n_{2}}] \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \boldsymbol{s}_{2}(t) \end{bmatrix}, \qquad (3)$$

where H_1 and H_2 are $m \times n_1$ and $m \times n_2$ mixing matrices corresponding to sources $s_1(t)$ and $s_2(t)$, respectively. Matrices H_1 and $[H_1 H_2]$ are assumed to be column full rank. **0** and **O** are a zero vector and a zero matrix, and the subscript shows their dimensions. The above two equations are mathematically equivalent but physically different. In the former way it is supposed that additional sources emit no signals, while in the latter way, additional sources have no paths to sensors. In this research, the latter case of Eq. (3) is considered because of mathematical convenience. In order to achieve BSS for $x_1(t)$, an $n_1 \times m$ separating matrix W_1 is estimated and the separated signals $y_1(t)$ is written as

$$y_1(t) = W_1 x_1(t) = P D s_1(t),$$
 (4)

where P is a permutation matrix and D is a diagonal matrix which results in a scaling factor. However, if $\hat{s}_1(t) = PDs_1(t)$ are considered as original sources, permutation and scaling ambiguity can be ignored.

After the appearance of new sources as shown in Fig. 1 (b), the mixed signals $\boldsymbol{x}_2(t)$ are written as follows:

$$\boldsymbol{x}_{2}(t) = [\boldsymbol{H}_{1} \ \boldsymbol{H}_{2}] \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \boldsymbol{s}_{2}(t) \end{bmatrix}.$$
(5)

Then a new $(n_1 + n_2) \times m$ matrix $W_2 = \begin{bmatrix} W_{21} \\ W_{22} \end{bmatrix}$ should be estimated. The goal of the BSS problem is to obtain the original sources $s_1(t)$ and $s_2(t)$ by estimating a separating matrix using ICA.

3 Separation Scheme

3.1 Separation of Initial Sources

In the first stage, we estimate a separating matrix for initially existing sources applying ICA. Without loss of generality, let us define an ideal matrix W_1 which is represented without permutation and scaling ambiguity as follows:

$$\boldsymbol{W}_1 = \boldsymbol{H}_1^{\dagger} = (\boldsymbol{H}_1^T \boldsymbol{H}_1)^{-1} \boldsymbol{H}_1^T, \tag{6}$$

where \dagger denotes the pseudo inverse (Moore-Penrose generalized inverse) operation.

Because H_1 is unknown, W_1 should be estimated by applying some independent component analysis (ICA) algorithm to the mixed signal $x_1(t)$. Inversely using estimated W_1 , H_1 can also be calculated as

$$\boldsymbol{H}_{1} = \boldsymbol{W}_{1}^{\dagger} = \boldsymbol{W}_{1}^{T} (\boldsymbol{W}_{1} \boldsymbol{W}_{1}^{T})^{-1}.$$
(7)

In this stage, we assume that we can estimate an ideal separating matrix, otherwise in the subsequent stage our method cannot avoid to fail the separation. Using the proposed method, the separation error in this stage affects the performance in the subsequent stage.

3.2 Separation After Sources Addition

When the appearance of the additional sources is detected, a new separating matrix should be calculated. As well as W_1 in (6), we can represent an ideal separating matrix $W_2 = \begin{bmatrix} W_{21} \\ W_{22} \end{bmatrix}$ as a function of H_1 and H_2 :

$$\boldsymbol{W}_{2} = [\boldsymbol{H}_{1} \ \boldsymbol{H}_{2}]^{\dagger} = \begin{bmatrix} \boldsymbol{H}_{1}^{\dagger} - \boldsymbol{H}_{1}^{\dagger} \boldsymbol{H}_{2} \left(\boldsymbol{H}_{2}^{T} (\boldsymbol{I}_{m} - \boldsymbol{H}_{1} \boldsymbol{H}_{1}^{\dagger}) \boldsymbol{H}_{2} \right)^{-1} \boldsymbol{H}_{2}^{T} (\boldsymbol{I}_{m} - \boldsymbol{H}_{1} \boldsymbol{H}_{1}^{\dagger}) \\ \left(\boldsymbol{H}_{2}^{T} (\boldsymbol{I}_{m} - \boldsymbol{H}_{1} \boldsymbol{H}_{1}^{\dagger}) \boldsymbol{H}_{2} \right)^{-1} \boldsymbol{H}_{2}^{T} (\boldsymbol{I}_{m} - \boldsymbol{H}_{1} \boldsymbol{H}_{1}^{\dagger}) \end{bmatrix}, \quad (8)$$

where I denotes an identity matrix with a subscript of the dimension. Note that the matrix $(I_m - H_1 H_1^{\dagger})$ is symmetric and idempotent, that is, $(I_m - H_1 H_1^{\dagger})$ is the orthogonal projection [6]. In the rest of this paper, $(I_m - H_1 H_1^{\dagger})$ is substituted with P_1 . Then the above equation can be rewritten in short form:

$$\boldsymbol{W}_{2} = \begin{bmatrix} \boldsymbol{W}_{21} \\ \boldsymbol{W}_{22} \end{bmatrix} = \begin{bmatrix} \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{I}_{m} - \boldsymbol{H}_{2} \left\{ \boldsymbol{P}_{1} \boldsymbol{H}_{2} \right\}^{\dagger} \right) \\ \left\{ \boldsymbol{P}_{1} \boldsymbol{H}_{2} \right\}^{\dagger} \end{bmatrix}$$
$$= \begin{bmatrix} \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{I}_{m} - \boldsymbol{H}_{2} \boldsymbol{W}_{22} \right) \\ \left\{ \boldsymbol{P}_{1} \boldsymbol{H}_{2} \right\}^{\dagger} \end{bmatrix}.$$
(9)

 W_2 is also obtained by applying ICA to the mixed signal $x_2(t)$. However, we may suffer a computational load, because we have to estimate a big separating matrix of size $(n_1 + n_2) \times m$. The essence of our method is to reuse the estimated mixing matrix H_1 and to reduce the cost. Here, let us define a global matrix $G = \begin{bmatrix} W_{21} \\ W_{22} \end{bmatrix} [H_1 \ H_2]$, which is shown in Fig. 2. If the submatrices of the global matrix satisfy the following conditions:

$$W_{21}H_1 = I_{n_1},$$
 (10)

$$W_{21}H_2 = O_{n_1 \times n_2},$$
 (11)

$$\boldsymbol{W}_{22}\boldsymbol{H}_1 = \boldsymbol{O}_{n_2 \times n_1},\tag{12}$$

(13)



Fig. 2. Global matrix after the new sources appearance

the separation problem can be simplified, because we only have to solve the separation of the part $W_{22}H_2$. The proposed scheme to estimate W_2 satisfying the above conditions is shown below.

Given H_2 , W_2 in (8) can be estimated because H_1 is known. However, there is no information about H_2 . Here, we substitute an arbitrary nonzero matrix A_2 for H_2 in W_2 of Eq. (8) and we obtain

$$\tilde{\boldsymbol{W}}_{2} = \begin{bmatrix} \tilde{\boldsymbol{W}}_{21} \\ \tilde{\boldsymbol{W}}_{22} \end{bmatrix} = \begin{bmatrix} \boldsymbol{H}_{1}^{\dagger} - \boldsymbol{H}_{1}^{\dagger}\boldsymbol{A}_{2} \left(\boldsymbol{A}_{2}^{T}\boldsymbol{P}_{1}\boldsymbol{A}_{2}\right)^{-1} \boldsymbol{A}_{2}^{T}\boldsymbol{P}_{1} \\ \left(\boldsymbol{A}_{2}^{T}\boldsymbol{P}_{1}\boldsymbol{A}_{2}\right)^{-1} \boldsymbol{A}_{2}^{T}\boldsymbol{P}_{1} \end{bmatrix}.$$
 (14)

Theorem 1. For any matrix A_2 , the following equation is satisfied,

$$W_{22}H_1 = O_{n_2 \times n_1}.$$
 (15)

 \tilde{W}_{22} works as a blocker [7] of H_1 even if the mixing matrix H_2 is unknown. Proof.

$$P_1H_1 = (I_m - H_1H_1^{\dagger})H_1$$

= $H_1 - H_1H_1^{\dagger}H_1$
= $H_1 - H_1$
= $O_{m \times n_1}$. (16)

Therefore $\tilde{\boldsymbol{W}}_{22}\boldsymbol{H}_1 = \left(\boldsymbol{A}_2^T\boldsymbol{P}_1\boldsymbol{A}_2\right)^{-1}\boldsymbol{A}_2^T\boldsymbol{P}_1\boldsymbol{H}_1 = \boldsymbol{O}_{n_2 \times n_1}.$

Using Theorem 1, it is clear that $\tilde{W}_{21}H_1 = I_{n_1}$.

Hereafter the method to estimated a separating matrix \hat{W}_{21} , which satisfies $\hat{W}_{21}H_2 = O_{n_1 \times M_2}$, is shown. For the purpose, the following equation is calculated:

$$\tilde{\boldsymbol{A}}_2 = \boldsymbol{C}_{x_2} \tilde{\boldsymbol{W}}_{22}^T, \tag{17}$$

where C_{x_2} is the covariance matrix of $x_2(t)$. The covariance matrix can be rewritten as

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$$\boldsymbol{C}_{x_2} = \left[\boldsymbol{H}_1 \ \boldsymbol{H}_2\right] \boldsymbol{C}_s \begin{bmatrix} \boldsymbol{H}_1^T \\ \boldsymbol{H}_2^T \end{bmatrix}, \qquad (18)$$

where C_s is the covariance matrix of s(t). Putting Eq. (18) into Eq. (17), we obtain

$$\tilde{\boldsymbol{A}}_{2} = [\boldsymbol{H}_{1} \ \boldsymbol{H}_{2}] \boldsymbol{C}_{s} \begin{bmatrix} \boldsymbol{H}_{1}^{T} \\ \boldsymbol{H}_{2}^{T} \end{bmatrix} \tilde{\boldsymbol{W}}_{22}^{T}$$

$$= [\boldsymbol{H}_{1} \ \boldsymbol{H}_{2}] \begin{bmatrix} \boldsymbol{C}_{s_{1}} & \boldsymbol{O}_{n_{1} \times n_{2}} \\ \boldsymbol{O}_{n_{2} \times n_{1}} & \boldsymbol{C}_{s_{2}} \end{bmatrix} \begin{bmatrix} \boldsymbol{O}_{n_{1} \times n_{2}} \\ \boldsymbol{H}_{2}^{T} \tilde{\boldsymbol{W}}_{22}^{T} \end{bmatrix}$$

$$= \boldsymbol{H}_{2} \boldsymbol{C}_{s_{2}} \boldsymbol{H}_{2}^{T} \tilde{\boldsymbol{W}}_{22}^{T}, \qquad (19)$$

where C_{s_1} and C_{s_2} are the covariance matrices of $s_1(t)$ and $s_2(t)$, respectively. For simplicity we replace $C_{s_2}H_2^T\tilde{W}_{22}^T$ by M. If det $M \neq 0$, that is, det $\tilde{W}_{22}H_2 \neq 0$, we substitute $\tilde{A}_2 = H_2M$ for H_2 of Eq. (9) to get $\hat{W}_2 = \begin{bmatrix} \hat{W}_{21} \\ \hat{W}_{22} \end{bmatrix}$,

$$\hat{\boldsymbol{W}}_{2} = \begin{bmatrix} \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{I}_{m} - \boldsymbol{H}_{2} \boldsymbol{M} \left\{ \boldsymbol{P}_{1} \boldsymbol{H}_{2} \boldsymbol{M} \right\}^{\dagger} \right) \\ \left\{ \boldsymbol{P}_{1} \boldsymbol{H}_{2} \boldsymbol{M} \right\}^{\dagger} \end{bmatrix}.$$
(20)

Here, we notice the following attractive property.

Lemma 1. $\{P_1H_2M\}^{\dagger}$ is a generalized inverse [8] of H_2M .

Proof. A generalized inverse of a matrix B, which is denoted as B^- , satisfies $BB^-B = B$. Thus $H_2M \{P_1H_2M\}^{\dagger} H_2M$ is calculated as follows:

$$H_2 M \{P_1 H_2 M\}^{\dagger} H_2 M$$

= $H_2 M \{M^T H_2^T P_1 H_2 M\}^{-1} M^T H_2^T P_1 H_2 M$
= $H_2 M.$ (21)

Therefore $\{P_1H_2M\}^{\dagger}$ is a generalized inverse of H_2M .

Then we have the following theorem.

Theorem 2. \hat{W}_{21} works as a blocker of H_1 , *i.e.*, $\hat{W}_{21}H_1 = O_{n_1 \times n_2}$. *Proof.* Using Lemma 1, it is clear that \hat{W}_{21} is a blocker of H_2 ;

$$\hat{\boldsymbol{W}}_{21}\boldsymbol{H}_{2} = \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{I}_{m} - \boldsymbol{H}_{2}\boldsymbol{M} \left\{ \boldsymbol{P}_{1}\boldsymbol{H}_{2}\boldsymbol{M} \right\}^{\dagger} \right) \boldsymbol{H}_{2}$$

$$= \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{H}_{2} - \boldsymbol{H}_{2}\boldsymbol{M} \left\{ \boldsymbol{P}_{1}\boldsymbol{H}_{2}\boldsymbol{M} \right\}^{\dagger} \boldsymbol{H}_{2}\boldsymbol{M}\boldsymbol{M}^{-1} \right)$$

$$= \boldsymbol{H}_{1}^{\dagger} \left(\boldsymbol{H}_{2} - \boldsymbol{H}_{2}\boldsymbol{M}\boldsymbol{M}^{-1} \right)$$

$$= \boldsymbol{O}_{n_{1} \times n_{2}}.$$
(22)

Table 1. Summary of the proposed method

	Step	Contents	
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1 Until $t < T_0$ given $\boldsymbol{x}_1(t)$, estimate \boldsymbol{W}_1 by ICA, and calculate $\boldsymbol{H}_1 = \boldsymbol{W}_1^{\dagger}$.

- 2 New sources appear at $t = T_0$.
- 3 Choose an arbitrary matrix A_2 and calculate \tilde{W}_2 in (14).
- 4 Calculate $\tilde{\boldsymbol{A}}_2 = \boldsymbol{C}_{x_2} \tilde{\boldsymbol{W}}_{22}^T$.
- 5 Substitute H_2 with \tilde{A}_2 and calculate \hat{W}_2 in (20).
- 6 Apply ICA to $\tilde{\boldsymbol{x}}_2(t) = \hat{\boldsymbol{W}}_{22} \boldsymbol{H}_2 \boldsymbol{s}_2(t)$.

Since \hat{W}_2 has the same property as \tilde{W}_2 , \hat{W}_{21} and \hat{W}_{22} work as blockers of H_2 and H_1 , respectively, so that the product of \hat{W}_2 and $x_2(t)$ is calculated as follows:

$$\hat{\boldsymbol{W}}_{2}\boldsymbol{x}_{2}(t) = \begin{bmatrix} \hat{\boldsymbol{W}}_{21}\boldsymbol{H}_{1} \ \hat{\boldsymbol{W}}_{21}\boldsymbol{H}_{2} \\ \hat{\boldsymbol{W}}_{22}\boldsymbol{H}_{1} \ \hat{\boldsymbol{W}}_{22}\boldsymbol{H}_{2} \end{bmatrix} \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \boldsymbol{s}_{2}(t) \end{bmatrix} \\
= \begin{bmatrix} \boldsymbol{I}_{n_{1}} & \boldsymbol{O}_{n_{1} \times n_{2}} \\ \boldsymbol{O}_{n_{2} \times n_{1}} \ \hat{\boldsymbol{W}}_{22}\boldsymbol{H}_{2} \end{bmatrix} \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \boldsymbol{s}_{2}(t) \end{bmatrix} \\
= \begin{bmatrix} \boldsymbol{s}_{1}(t) \\ \tilde{\boldsymbol{x}}_{2}(t) \end{bmatrix}.$$
(23)

As a result, the BSS problem of $\tilde{x}_2(t) = \hat{W}_{22}H_2s_2(t)$ has only to be solved. Our method avoids to calculate the whole separating matrix W_2 by ICA. It implies a low computational cost. The proposed method is summarized in Table 1.

4 Simulation

For the sake of clarity of the proposed method, BSS computational simulations were conducted on a PC (Core2 Duo E4300, 1.0GB RAM) using FastICA algorithm [9] implemented on MATLAB. Speech signals were used as source signals. The number of initial sources n_1 was four and the number of additional sources n_2 was varied from one to three. The number of sensors m was seven and the mixed signals were generated through a mixing matrix. The coefficients of the mixing matrix were set randomly. We measured the processing time and the separation performance in order to evaluate the proposed method and the conventional method, which estimates a new separating matrix without any initialization. As a performance measure, signal to interference ratio (SIR) was calculated as follows:

$$SIR = \frac{1}{n_1 + n_2} \sum_{i=1}^{n_1 + n_2} 10 \log_{10} \frac{\sum_t y_{i,s_j}^2(t)}{\sum_{k \neq j} \sum_t y_{i,s_k}^2(t)},$$
(24)

where $y_{i,s_j}(t)$ and $y_{i,s_k}(t)$ are *i*th separated signals which are contributed by a certain source $s_j(t)$ and an interfering source $s_k(t)$, respectively. The results of 100 trials are shown. Fig. 3 (a) illustrates the processing time varying the number of the new sources. It is clear that the processing time using the proposed



Fig. 3. The performance of the proposed method and the conventional method varying the number of additional sources. (a) The averaged processing time with standard deviation. (b) The averaged SIR with standard deviation.

method is significantly smaller than the conventional one. Our concern lies in the fact to estimate as $n_2 \times n_2$ separating matrix, instead of the conventional method that estimates $(n_1 + n_2) \times m$ separating matrix. Fig. 3 (b) depicts the separation performance. It is shown that our method outperforms the conventional method since the blockers work efficiently. Furthermore the conventional method caused permutation problems due to re-estimation of the separating matrix. The output channels of initial sources were different from those obtained by initial matrix, while using the proposed method the initial sources $s_1(t)$ are given in the same order despite the appearance of new sources.

5 Discussion

The problem of detecting the number of active sources is very important for BSS and have been discussed by many authors [10,11,12]. Although some problems

need to be solved, for instance when we have additional sources that appear after estimation of the separating matrix. In this research we proposed a new method that is based on the number of sources known a priori and the number of the dominant eigenvalues of the covariance matrix that can be solved straightforward using the well known FastICA algorithm. The highlight in this problem is to determine the sources dimension that is not solved easily by any BSS algorithm.

Until now we discussed the separation in the case when new sources appear. In addidion, the separation with a disappearance can be solved by a simple procedure based on the columns of the mixing matrix. Moreover, if a disappearance is detected the output energy become zero, which make the algorithm identify the vanished sources. Consequently, the corresponding columns of the vanished sources are removed from the mixing matrix, which was estimated before the disappearance, resulting in a new mixing matrix. Finally a new separating matrix is given as a pseudo inverse of the mixing matrix.

6 Conclusion and Future Works

In this manuscript we have proposed a scheme to separate mixed signals in the situation of new sources appearing. Utilizing the covariance matrix of the mixed signals, blockers of the initial sources and the additional sources can be obtained. As a result, blind separation of the new sources is only needed. Thus the computational load can be reduced compared with the conventional method. Through a computer simulation validity of the proposed method is shown.

Assuming that the matrix $\boldsymbol{M} = \boldsymbol{C}_{s_2} \boldsymbol{H}_2^T \tilde{\boldsymbol{W}}_{22}^T$ is nonsingular, namely $\tilde{\boldsymbol{W}}_{22} \boldsymbol{H}_2$ is also regular. To obtain $\tilde{\boldsymbol{W}}_{22}$ an arbitrary matrix \boldsymbol{A}_2 is used, further investigation about the way of choosing \boldsymbol{A}_2 is needed such that $\tilde{\boldsymbol{W}}_{22} \boldsymbol{H}_2 = \{\boldsymbol{P}_1 \boldsymbol{A}_2\}^{\dagger} \boldsymbol{H}_2$ is full rank.

The next goal is the extension of the proposed method to convolutive mixture to achieve blind separation of audio signals.

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